

CircuitTSAT

A Solver for Large Instances of the Disjunctive Temporal Problem

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Introduction

- We study fast solution to the Disjunctive Temporal Problem (DTP)
- We present CircuitTSAT that uses circuit-based approaches similar to [Bryant et al, 2007] for DTPs.
 - We show that structure of the DTP is amenable to circuit representations
- We explore how our approach compares with other approaches to other DTP solvers and discuss the advantages and disadvantages of CircuitTSAT

Motivating Example

Scheduling

- Consider set of N events $\{X_i\}$
- The i -th event has duration d_i
- Some events *must* occur before others:

$$X_i \prec X_j$$

- Single Resources: force events to be disjoint

$$X_i \prec X_j \vee X_j \prec X_i$$

Motivating Example

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$$X_j - X_i \geq d_i$$

A simple temporal constraint!

- Single Resources: force events to be disjoint

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A simple temporal constraint!

- Single Resources: force events to be disjoint

$$X_j - X_i \geq d_i \quad \vee \quad X_i - X_j \geq d_j$$

A disjunction of simple temporal constraints

Simple Temporal Problems

- An STP is characterized by
 - 1 A set of events: $\mathcal{X} = \{X_0, X_1 \dots X_N\}$
 - 2 A set of simple temporal constraints (STCs) of the form

$$LB(i, j) \leq X_j - X_i \leq UB(i, j)$$

- A solution is an assignment \mathcal{X} satisfying all STCs
- STPs can be efficiently solved using shortest path algorithms [Dechter et al, 1991]

Disjunctive Temporal Problems

- A DTP is characterized by
 - 1 Events: $\mathcal{X} = \{X_0, X_1 \dots X_N\}$
 - 2 Disjunctive temporal constraints (DTCs) of the form

$$\mathcal{S}_1(\mathbf{z}_1) \vee \mathcal{S}_2(\mathbf{z}_2) \vee \dots \vee \mathcal{S}_K(\mathbf{z}_K)$$

where $\mathcal{S}_k(\mathbf{z}_k)$ is an STC of the form

$$X_{i_k} - X_{j_k} \leq z_k$$

for a pair of temporal variables

- A solution is an assignment \mathcal{X} satisfies every DTCs.
- Solving DTPs are NP-hard in general.

Previous Work

- **TSAT++** [Armando et al, 2000, Armando et al, 2004]
 - Each STC is represented by a propositional variable
 - Satisfying the resulting SAT instance yields an STP
 - Consistency of the induced STP finds DTP solutions
- **Satisfiability Modulo Theories (SMT) Solvers**
 - SMT solvers generalize SAT instances
 - Domain-specific functions replace Boolean variables
 - DTPs can be cast as SMT(DL)
- **UCLID** [Bryant et al, 2002, Bryant et al, 2007]
 - UCLID SMT solver employs circuit-based expansions
 - Modern SMT solvers outperform circuit-based approaches
 - We re-explore circuit-based approaches for DTPs

CircuitTSAT

A circuit-based approach to DTPs

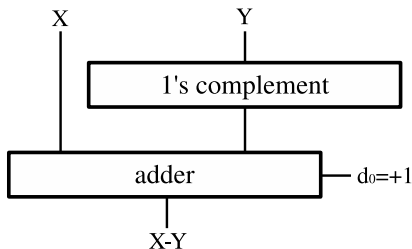
Representing STCs as Circuits

- Let X and Y be integral variables.
- Further, suppose both are between 0 and Q .
- Then both can be represented by $q = \lceil \log_2 Q \rceil$ bits:

$$X = \langle x_q, x_{q-1} \dots x_1 \rangle$$

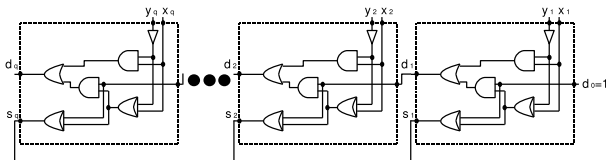
$$Y = \langle y_q, y_{q-1} \dots y_1 \rangle$$

- $X - Y$ can then be represented as a circuit:



Representing STCs as Circuits

Sum and Carry Bits



- The $X - Y$ difference circuit produces sum S and carry D bits:

$$S = \langle s_q, s_{q-1} \dots s_1 \rangle$$

$$D = \langle d_q, d_{q-1} \dots d_1, d_0 = 1 \rangle$$

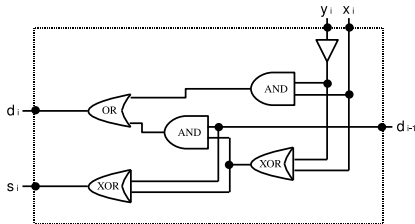
- Since $X \geq 0$ and $Y \geq 0$, $X - Y$ cannot overflow.
- Further, d_q indicates the sign of $X - Y$:

$$d_q \Leftrightarrow X - Y \geq 0$$

Representing STCs as Circuits

Auxiliary Variables for $X - Y \geq 0$

Consider a Full Adder Circuit



We construct auxiliary variables d_i

$$\begin{aligned} d_i &\Leftrightarrow [(x_i \oplus \bar{y}_i) \wedge d_{i-1}] \vee (x_i \wedge \bar{y}_i) \\ &\Leftrightarrow [(x_i \vee \bar{y}_i) \wedge d_{i-1}] \vee (x_i \wedge \bar{y}_i) \end{aligned}$$

Representing STCs as Circuits

Extending to $Y - X \leq z$

- Constraints of $X - Y \geq 0$ extend to $Y - X \leq z$.
- By regrouping, we convert it to 2 sums:

$$(X - Y) + z \geq 0$$

Representing STCs as Circuits

Extending to $Y - X \leq z$

- Constraints of $X - Y \geq 0$ extend to $Y - X \leq z$.
- By regrouping, we convert it to 2 sums:

$$\underbrace{(X - Y)}_{1^{\text{st}} \text{ term} = S} + z \geq 0$$

- The 1st is the difference between two variables: S .

Representing STCs as Circuits

Extending to $Y - X \leq z$

- Constraints of $X - Y \geq 0$ extend to $Y - X \leq z$.
- By regrouping, we convert it to 2 sums:

$$\underbrace{S + z}_{2^{\text{nd}} \text{ term}} \geq 0$$

- The 2^{nd} adds a constant to that difference.

Representing STCs as Circuits

Extending to $Y - X \leq z$

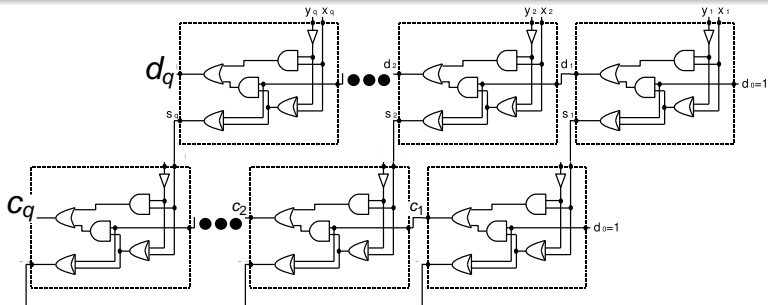
- Constraints of $X - Y \geq 0$ extend to $Y - X \leq z$.
- By regrouping, we convert it to 2 sums:

$$S + z \geq 0$$

- We need a circuit for $S + z \geq 0$

Representing STCs as Circuits

Extending to $S + z \geq 0$



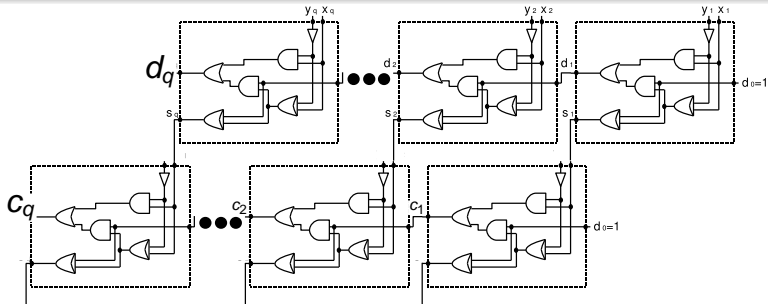
- Terms of S are $s_i \Leftrightarrow d_{i-1} \oplus (x_i \oplus \bar{y}_i)$
- Carry bits of $S + z$ compiled from bits of z

$$c_i \Leftrightarrow [(s_i \vee \textcircled{Z}_i) \wedge c_{i-1}] \vee (s_i \wedge \textcircled{Z}_i)$$

where \textcircled{Z}_i depends on sign of z

Representing STCs as Circuits

Extending to $S + z \geq 0$



- Terms of S are $s_i \Leftrightarrow d_{i-1} \oplus (x_i \oplus \bar{y}_i)$
- Carry bits of $S + z$ compiled from bits of z

$$c_i \Leftrightarrow \begin{cases} c_{i-1} \vee s_i & \text{if } \textcircled{z}_i = 1 \\ c_{i-1} \wedge s_i & \text{if } \textcircled{z}_i = 0 \end{cases}$$

where \textcircled{z}_i depends on sign of z

Representing STCs as Circuits

Extending to $S + z \geq 0$

- $X - Y + z \geq 0$ expressed by c_q but it can overflow
- There are 2 cases:
 - 1 $X - Y$ and z agree in sign
 - Adding z to $X - Y$ will not change its sign
 - Hence d_q indicates the sign of $X - Y + z$
 - 2 $X - Y$ and z have opposite sign
 - Adding z to $X - Y$ cannot overflow
 - Hence c_q indicates the sign of $X - Y + z$
- We can represent the DTCs state by
 - $d_q \wedge c_q$ if $z_w \leq 0$
 - $d_q \vee c_q$ if $z_w > 0$

Representing DTCs Propositionally

- DTCs are disjunctions of STCs.
- The w^{th} STC $Y - X \leq z$ can be represented propositionally by a_w :

$$a_w \Leftrightarrow d_{q,w} \wedge c_{q,w} \quad \text{if } z_w \leq 0$$

$$a_w \Leftrightarrow d_{q,w} \vee c_{q,w} \quad \text{if } z_w > 0$$

- The DTC can then be expressed as a CNF clause:

$$a_1 \vee a_2 \vee \dots \vee a_k$$

CircuitTSAT: Representing DTPs as CNF

- A DTP is expressed in CNF by difference logic:
 - Carry bit d_q of $X - Y$ is expressed in CNF
 - Carry bit c_q of $X - Y + z$ is expressed in CNF
 - w^{th} STC $Y - X \leq z$ is expressed in CNF in terms of $c_{q,w}$ and $d_{q,w}$ as a_w .
 - Each DTC is represented in CNF
- Solutions to the DTP are found using modern SAT Solver. We tested several solvers:
 - JeruSAT [Nadel, 2002]
 - MiniSAT [Eén and Sörensson, 2003]
 - zChaff

Trade-offs with CircuitTSAT

- Size of CNF (For a DTP with N temporal variables, M clauses, and K disjuncts per clause)
 - $O(qN + qKM)$ propositional variables
 - $O(qKM)$ clauses
- Choosing the size of the bit-space
 - Worst case: Chain of largest constant from each DTC
 - In a chain, these constants sum

$$q_{max} = \left\lceil \log_2 \sum_{i=1}^M \max_{z_w \in DTC_i} |z_w| \right\rceil$$

- Smaller q gives faster performance, but is incomplete
- Non-integer problems
 - Negative variables are avoided by translation via X_0
 - Real-valued variables approximated via scaling

Experiments

Experimental Setup

- We compare DTP solvers on random DTPs
- Experimental parameters
 - N — number of temporal variables
 - M/N — ratio of clauses to temporal variables
 - K — number of disjuncts per DTC
 - L — maximum magnitude of any constant in the DTP
- Algorithms we compare
 - CircuitTSAT
 - TSAT++ [Armando et al, 2000, Armando et al, 2004]
 - Yices [Dutertre and de Moura, 2006]

Experiment 1

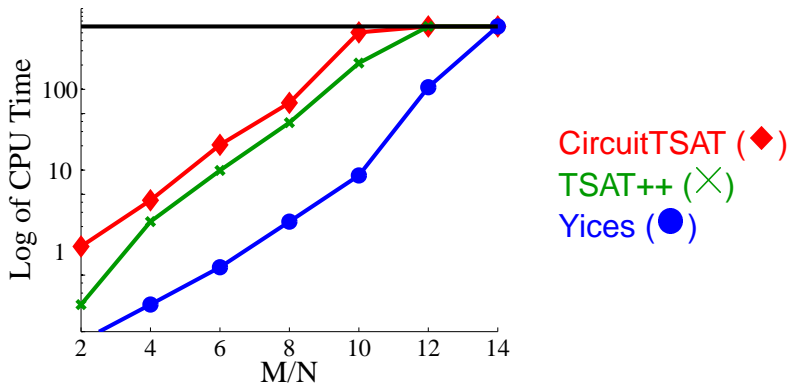
Comparing Performance for an Increasing Number of Literals

- We test for large numbers of temporal variables (N) and many disjuncts per clause (K)
- We plot log exec time vs. clause to variable ratio (M/N)

Performance Comparison of CircuitTSAT

Increasing Number of Literals

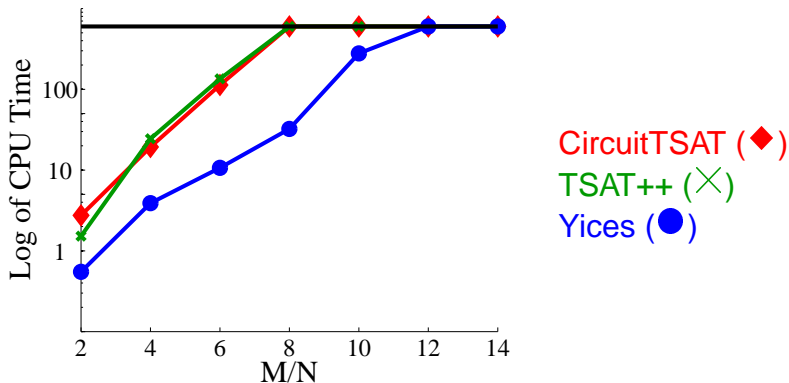
Logarithm of Median Running Times for
 $K = 3$ and $N = 50$



Performance Comparison of CircuitTSAT

Increasing Number of Literals

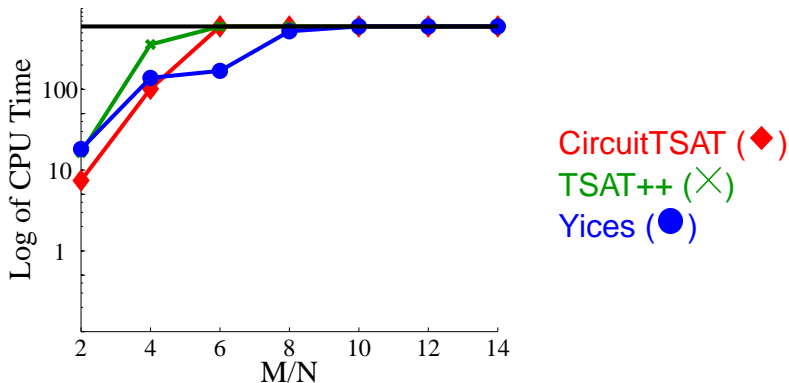
Logarithm of Median Running Times for
 $K = 3$ and $N = 100$



Performance Comparison of CircuitTSAT

Increasing Number of Literals

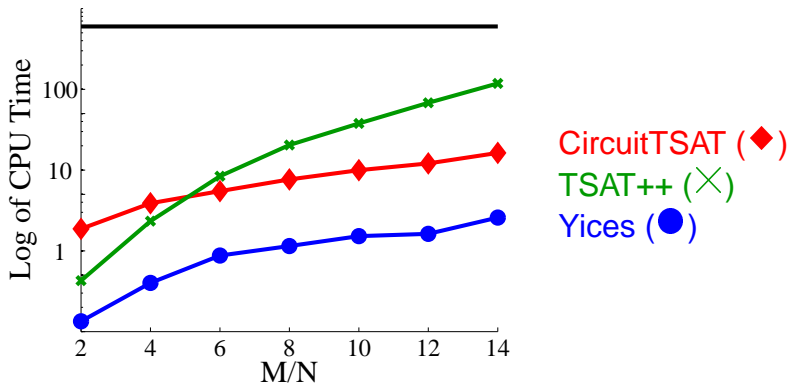
Logarithm of Median Running Times for
 $K = 3$ and $N = 200$



Performance Comparison of CircuitTSAT

Increasing Number of Literals

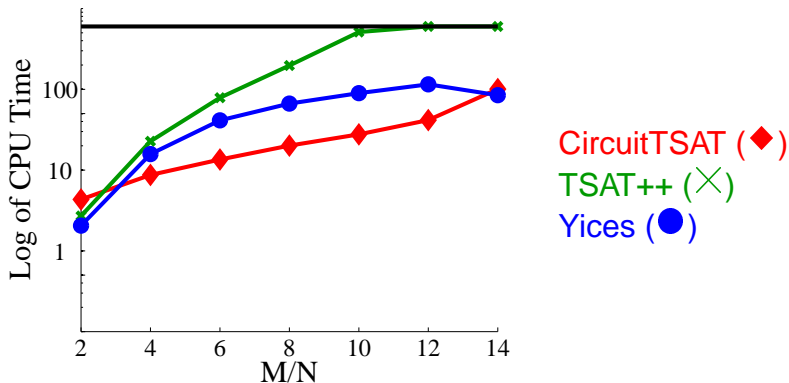
Logarithm of Median Running Times for
 $K = 5$ and $N = 50$



Performance Comparison of CircuitTSAT

Increasing Number of Literals

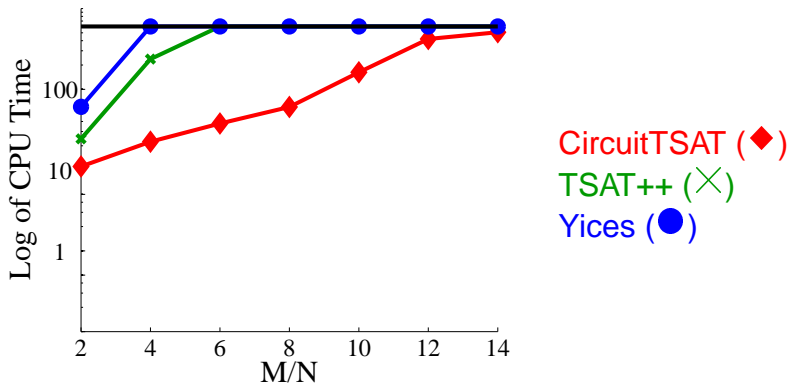
Logarithm of Median Running Times for
 $K = 5$ and $N = 100$



Performance Comparison of CircuitTSAT

Increasing Number of Literals

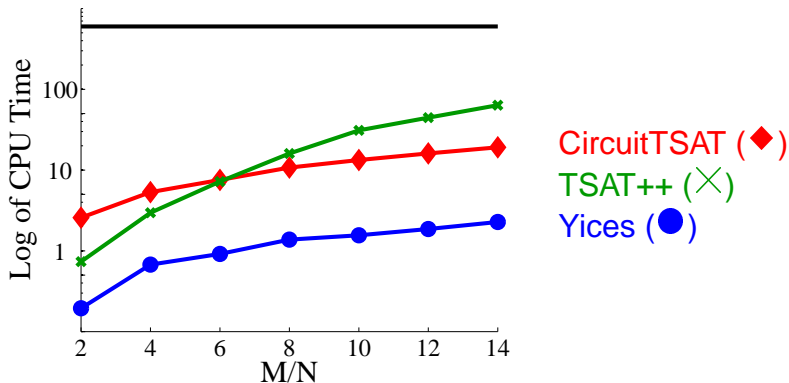
Logarithm of Median Running Times for
 $K = 5$ and $N = 200$



Performance Comparison of CircuitTSAT

Increasing Number of Literals

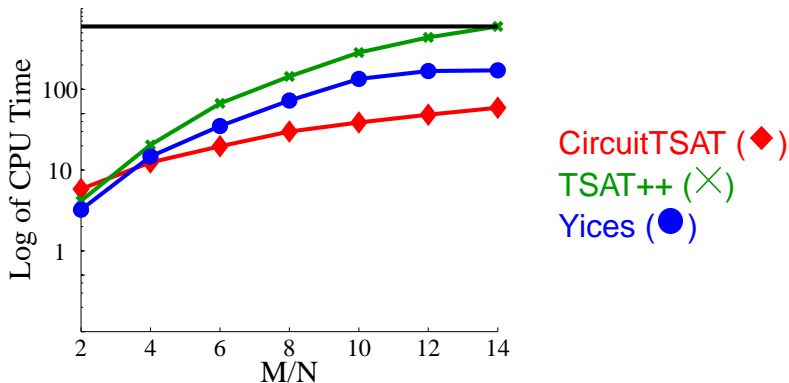
Logarithm of Median Running Times for
 $K = 7$ and $N = 50$



Performance Comparison of CircuitTSAT

Increasing Number of Literals

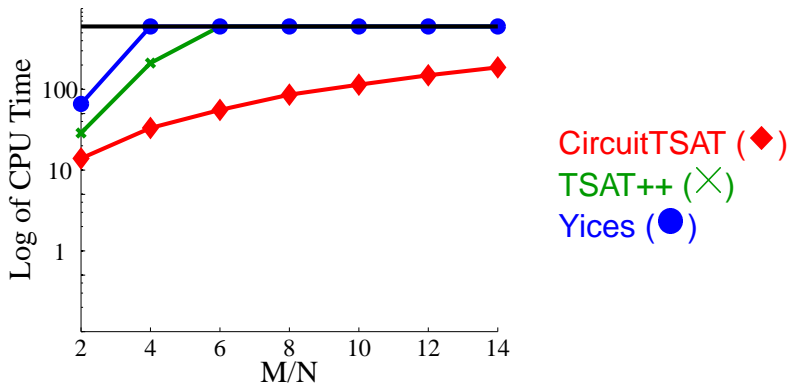
Logarithm of Median Running Times for
 $K = 7$ and $N = 100$



Performance Comparison of CircuitTSAT

Increasing Number of Literals

Logarithm of Median Running Times for
 $K = 7$ and $N = 200$



Future Work

Extending Circuit-based Approaches to DTPs

- Preprocessing steps similar to TSAT++
- Better strategies for allocating bits
 - Using q_{max} bits is a conservative allocation approach
 - May be able to allocate each X_i with different size
 - May be able to employ over-approximation of [Bryant et al, 2007]
- Expanding to Hybrid CSPs
 - Extending CircuitTSAT to hybrid constraints
 - Applications to more problems in planning and scheduling
- Better SAT solvers
 - SAT solvers could be designed to better exploit the structural information from a DTP
 - Similarly the MODOC solver was designed for planning [Gelder and Okushi, 1999]

Conclusion

- We demonstrated the capabilities of CircuitTSAT — a circuit-based approach for solving DTP instances
- We compared the circuit-based approach to other DTP and SMT solvers
 - Yices faster for smaller values of K and N
 - For large K and N , CircuitTSAT scales better and significantly outperformed both Yices and TSAT++
 - CircuitTSAT exploits structural information of DTPs
- It remains to be seen if circuit-based approaches can be integrated with other approaches for better overall performance

The End
Any question?

Experiment 2

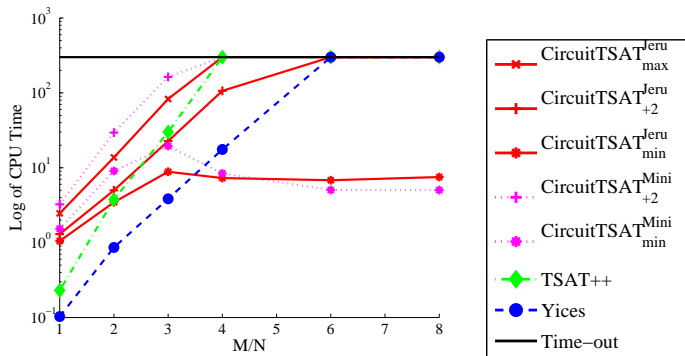
Comparing Performance for Different Variants of TSAT

- Here we compare the execution times of variants of CircuitTSAT
- We explore performance for large numbers of temporal variables (N) and many disjuncts per clause (K)
- In the following plots, we plot log running time vs. clause to variable ratio (M/N)
- These tests demonstrate the effectiveness of different SAT solvers for our approach and the advantages of using a smaller bit-space size q .

Performance Comparison of CircuitTSAT

Variants of CircuitTSAT

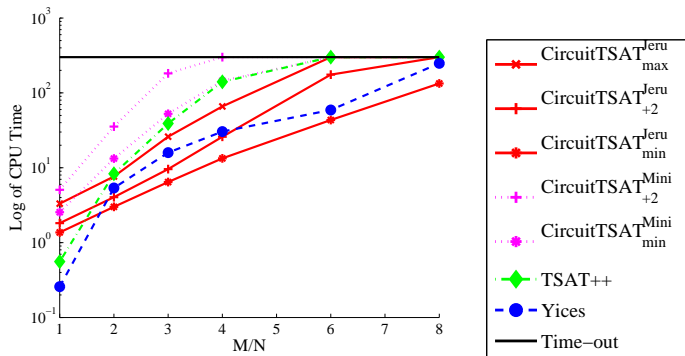
Logarithm of Median Running Times for
 $N = 150$ and $K = 2$



Performance Comparison of CircuitTSAT

Variants of CircuitTSAT

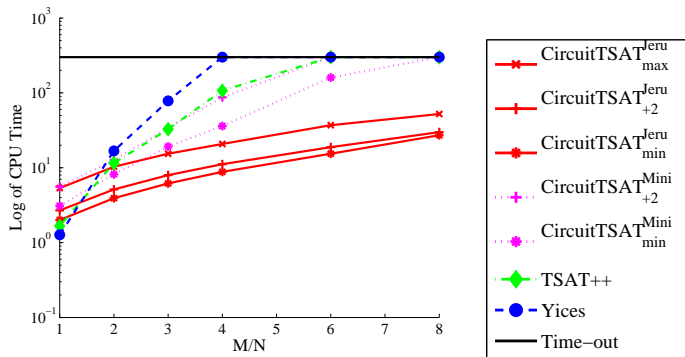
Logarithm of Median Running Times for
 $N = 150$ and $K = 3$



Performance Comparison of CircuitTSAT

Variants of CircuitTSAT

Logarithm of Median Running Times for
 $N = 150$ and $K = 5$



Extra Slides

DTP Notation

- DTC \mathcal{D} and a DTP \mathcal{P} then expressed as follows:

$$\mathcal{D}_{\Lambda_\ell} = \bigvee_{(\lambda_i, \gamma_i, \mathbf{z}_i) \in \Lambda_\ell} \mathcal{S}_{\gamma_i, \lambda_i}(\mathbf{z}_i) ; \quad \mathcal{P} = \bigwedge_{\ell=1}^M \mathcal{D}_{\Lambda_\ell}$$

where Λ_ℓ is a set containing triplets $(\lambda_i, \gamma_i, \mathbf{z}_i)$

Extra Slides

Handling Overflows

- $X - Y + z \geq 0$ is expressed by c_q and d_q but it can overflow
- There are 2 cases:
 - ① $z \leq 0$:
 - $d_q = 0$ implies that $X - Y < 0$ hence $\bar{d}_q \Rightarrow X - Y + z < 0$
 - $d_q = 1$ implies that $X - Y \geq 0$ so an overflow can't occur. Hence, $c_q \Leftrightarrow X - Y + z \geq 0$.
$$(X - Y + z \geq 0) \Leftrightarrow (d_q \wedge c_q)$$
 - ② $z > 0$:
 - $d_q = 1$ implies that $X - Y \geq 0$ hence $d_q \Rightarrow X - Y + z \geq 0$.
 - $d_q = 0$ implies that $X - Y < 0$ so an overflow can't occur. Hence, $c_q \Leftrightarrow X - Y + z \geq 0$.
$$(X - Y + z \geq 0) \Leftrightarrow (d_q \vee c_q)$$
- We can represent the DTCs state by $d_q \wedge c_q$ or

Extra Slides

CNF for d_i carry bits

- The CNF for d_i is

$$\begin{array}{llll} (d_i \vee d_{i-1} \vee \bar{x}_i \vee y_i) & \wedge & (d_i \vee \bar{d}_{i-1} \vee y_i) & \wedge \\ (d_i \vee \bar{d}_{i-1} \vee \bar{x}_i) & \wedge & (\bar{d}_i \vee \bar{d}_{i-1} \vee x_i \vee \bar{y}_i) & \wedge \\ (\bar{d}_i \vee d_{i-1} \vee \bar{y}_i) & \wedge & (\bar{d}_i \vee d_{i-1} \vee x_i) & \end{array}$$

Extra Slides




CNF for c_i carry bits

- The CNF for c_i is

condition A	condition B
$c_i \vee \bar{c}_{i-1}$	$\bar{c}_i \vee c_{i-1}$
$c_i \vee c_{i-1} \vee d_{i-1} \vee x_i \vee y_i$	$c_i \vee \bar{c}_{i-1} \vee d_{i-1} \vee x_i \vee y_i$
$c_i \vee c_{i-1} \vee d_{i-1} \vee \bar{x}_i \vee \bar{y}_i$	$c_i \vee \bar{c}_{i-1} \vee d_{i-1} \vee \bar{x}_i \vee \bar{y}_i$
$c_i \vee c_{i-1} \vee \bar{d}_{i-1} \vee \bar{x}_i \vee y_i$	$c_i \vee \bar{c}_{i-1} \vee \bar{d}_{i-1} \vee \bar{x}_i \vee y_i$
$c_i \vee c_{i-1} \vee \bar{d}_{i-1} \vee x_i \vee \bar{y}_i$	$c_i \vee \bar{c}_{i-1} \vee \bar{d}_{i-1} \vee x_i \vee \bar{y}_i$
$\bar{c}_i \vee c_{i-1} \vee d_{i-1} \vee \bar{x}_i \vee y_i$	$\bar{c}_i \vee \bar{c}_{i-1} \vee d_{i-1} \vee \bar{x}_i \vee y_i$
$\bar{c}_i \vee c_{i-1} \vee d_{i-1} \vee x_i \vee \bar{y}_i$	$\bar{c}_i \vee \bar{c}_{i-1} \vee d_{i-1} \vee x_i \vee \bar{y}_i$
$\bar{c}_i \vee c_{i-1} \vee \bar{d}_{i-1} \vee x_i \vee y_i$	$\bar{c}_i \vee \bar{c}_{i-1} \vee \bar{d}_{i-1} \vee x_i \vee y_i$
$\bar{c}_i \vee c_{i-1} \vee \bar{d}_{i-1} \vee \bar{x}_i \vee \bar{y}_i$	$\bar{c}_i \vee \bar{c}_{i-1} \vee \bar{d}_{i-1} \vee \bar{x}_i \vee \bar{y}_i$

Extra Slides

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